

# Signals and Systems I

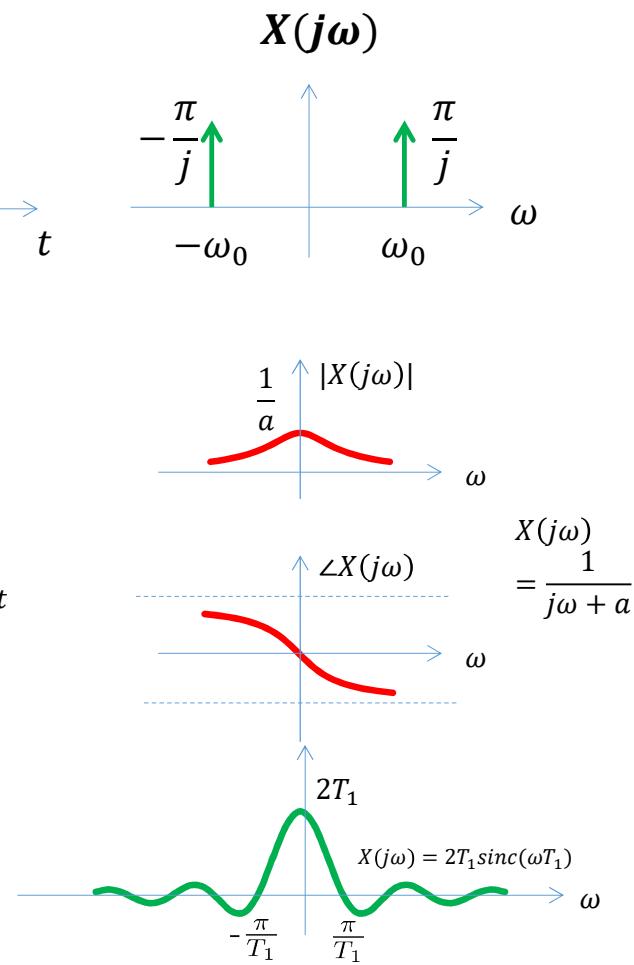
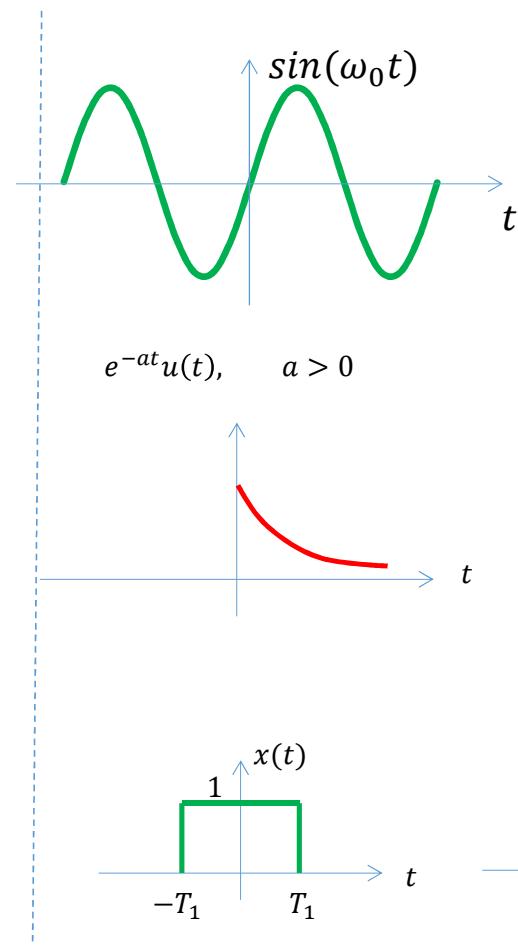
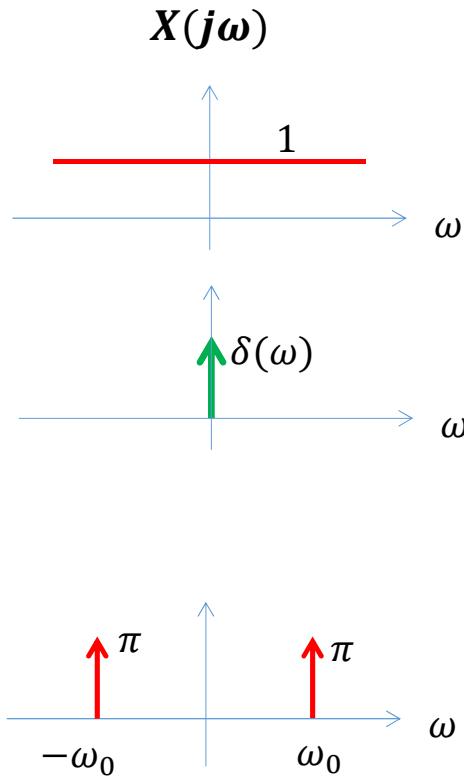
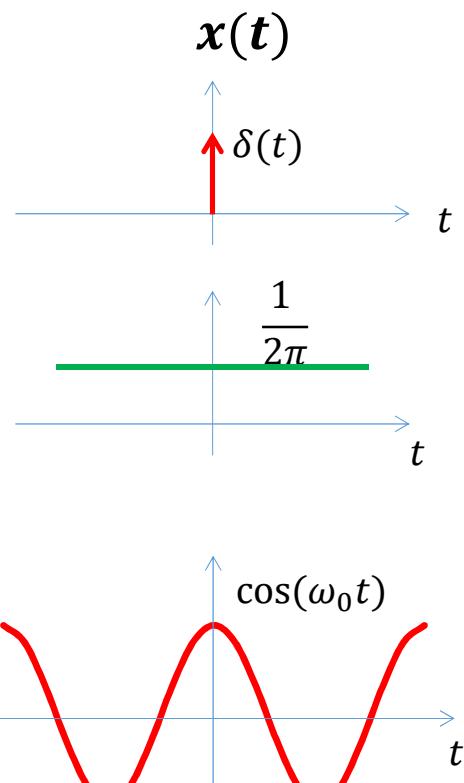
Topic 10

$$\text{Synthesis Equation: } x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

$$\text{Analysis Equation: } X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

- More on calculation of FT and additional FT properties
- FT & LTI systems
- FT & LTIDE Systems
- Modulation & Demodulation
- Sampling

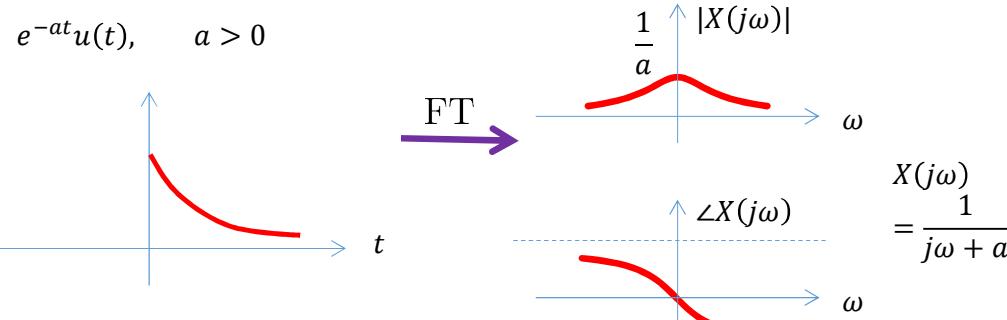
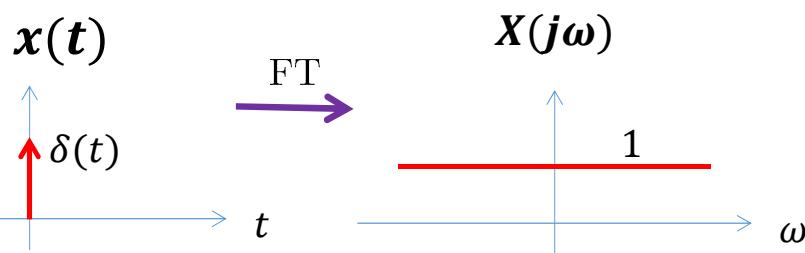
## Fourier Transform of some Important Signals (review)



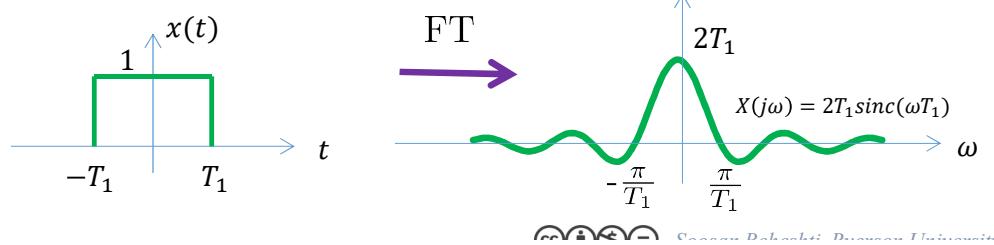
# Fourier Transform of some Important Signals

Synthesis Equation:  $x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$

Analysis Equation:  $X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$

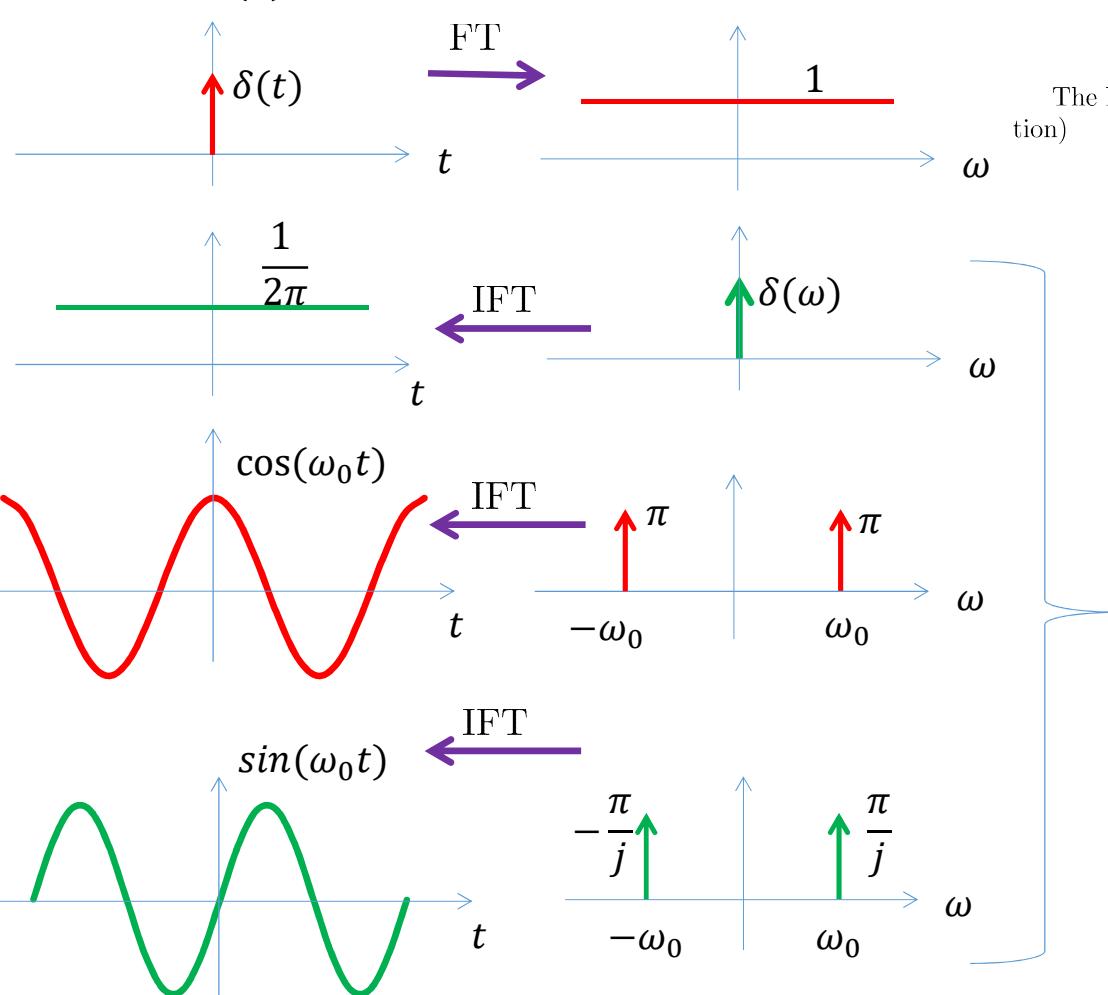


The FTs can be calculated using the FT equation (The Analysis equation)



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## Fourier Transform of some Important Signals



Synthesis Equation:  $x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$

Analysis Equation:  $X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$

The FT can be easily calculated using the FT equation (The Analysis equation)

The inverse FT (IFT) can be easily calculated using:

1-The IFT equation (Synthesis equation)

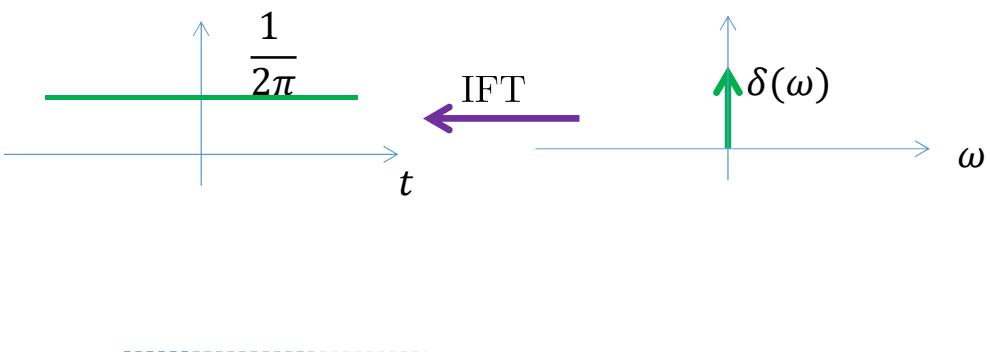
2- Duality property

The FT can be calculated

3- using the periodic property

4- or directly by using the synthesis equation! (not as convenient as the above methods)

# Fourier Transform of a constant signal



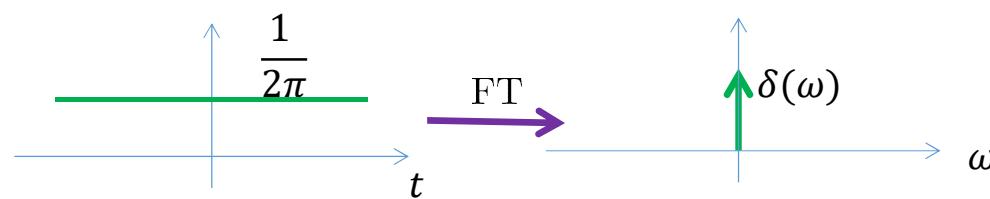
The inverse FT (IFT) can be easily calculated using:

1-The IFT equation (Synthesis equation)

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} \delta(\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} \delta(\omega) d\omega = \frac{1}{2\pi}$$

2- Duality property

$$\delta(t) \xrightarrow{\text{FT}} \Delta(j\omega) = 1 \xrightarrow{\text{IFT}} 2\pi\delta(\omega)$$



The FT can be calculated

3- using the periodic property:

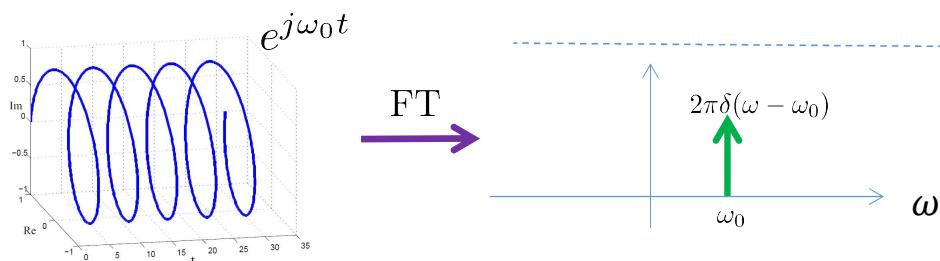
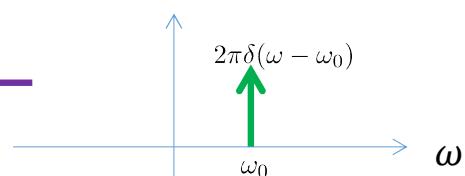
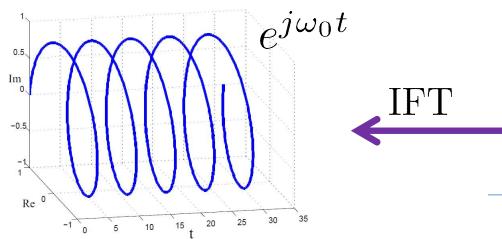
Constant signal is a pulse from  $-T$  to  $T$  with period  $2T$ :

$$D_0 = \frac{1}{2\pi} \text{ and the rest of the } D_n \text{s are zero. Therefore } X(j\omega) = 2\pi \times D_0\delta(\omega)$$

4- or directly by using the synthesis equation! (not as convenient as the above methods):

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt = \int_{-\infty}^{\infty} \frac{1}{2\pi} e^{-j\omega t} dt = \begin{cases} \delta(\omega) & \text{if } \omega = 0 \\ 0 & \text{if } \omega \neq 0 \end{cases} \text{ (why?)}$$

# Fourier Transform of Periodic Spiral



Orthogonality principle:

$$\int_{-\infty}^{\infty} e^{j\omega_0 t} e^{j\omega t} dt = \begin{cases} 0 & \text{if } \omega_0 \neq \omega \\ 2\pi\delta(\omega - \omega_0) & \text{if } \omega_0 = \omega \end{cases}$$

Remember:

$$e^{j\omega_0 t} \xrightarrow{\text{Fourier Transform}} 2\pi\delta(\omega - \omega_0)$$

The inverse FT (IFT) can be easily calculated using:

1- Direct use of IFT definition:

$$\begin{aligned} x(t) &= \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} 2\pi\delta(\omega - \omega_0) e^{j\omega t} d\omega \\ &= \int_{-\infty}^{\infty} \delta(\omega - \omega_0) e^{j\omega_0 t} d\omega = e^{j\omega_0 t} \underbrace{\int_{-\infty}^{\infty} \delta(\omega - \omega_0) d\omega}_{1} = e^{j\omega_0 t} \end{aligned}$$

2- Frequency Shift property

$$\delta(\omega) \xrightarrow{\text{IFT}} \frac{1}{2\pi} (\text{constant signal})$$

Shifting the FT by  $\omega$  is multiplication by  $e^{j\omega_0 t}$  in time.

3- Shift in time property and Duality property

$$x_1(t) = \delta(t + \omega_0) \xrightarrow{\text{FT}} X(j\omega) = e^{j\omega_0 \omega}$$

$$x_2(t) = X(jt) = e^{j\omega_0 t} \xrightarrow{\text{FT}} 2\pi x_1(-\omega) = 2\pi\delta(\omega - \omega_0)$$

The FT can be calculated

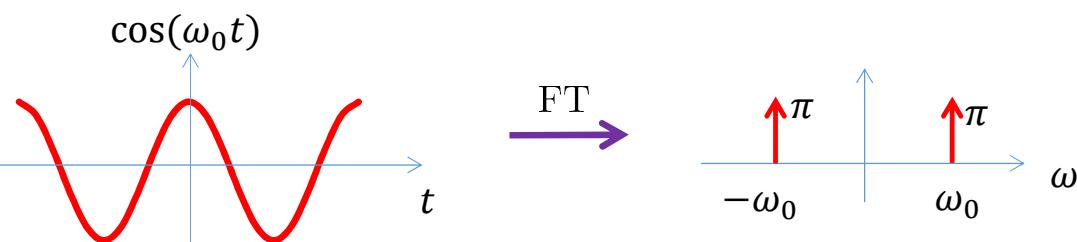
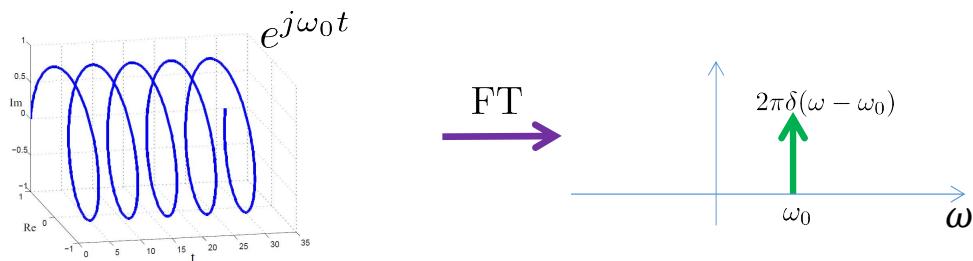
4- using the periodic property:

It is known that periodic spiral has only one nonzero  $D_n$ . If  $\omega$  is positive  $D_1 = 1$  and if it is negative  $D_{-1} = 1$ . Therefore  $X(j\omega) = 2\pi \times D_1 \delta(\omega - \omega_0)$  for positive  $\omega_0$  and  $X(j\omega) = 2\pi \times D_{-1} \delta(\omega - \omega_0)$  for negative  $\omega_0$ .

5- or directly by using the synthesis equation! (not as convenient as the above methods):

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt = \int_{-\infty}^{\infty} e^{j\omega_0 t} e^{-j\omega t} dt = 2\pi\delta(\omega - \omega_0)$$

# Fourier Transform of Cosine



Using either of five methods explained in the previous page, FT of cosine signal can be found through linear property of FT:

$$\begin{aligned}
 x(t) = \cos(t) &= \frac{1}{2}e^{j\omega_0 t} + \frac{1}{2}e^{-j\omega_0 t} \xrightarrow{\text{FT}} X(j\omega) = \frac{1}{2}\text{FT}(e^{j\omega_0 t}) + \frac{1}{2}\text{FT}(e^{-j\omega_0 t}) \\
 &= \frac{1}{2}2\pi\delta(\omega - \omega_0) + \frac{1}{2}2\pi\delta(\omega + \omega_0)
 \end{aligned}$$

## Fourier Transform Properties (Product of Signals)

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$
$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

Remember convolution in time

$$x_1(t) \xrightarrow{FT} X_1(j\omega)$$
$$x_2(t) \xrightarrow{FT} X_2(j\omega)$$
$$x_3(t) = x_1(t) * x_2(t) \xrightarrow{FT} X_3(j\omega) = X_1(j\omega) \times X_2(j\omega)$$

Convolution in time  $\equiv$  Product in Frequency

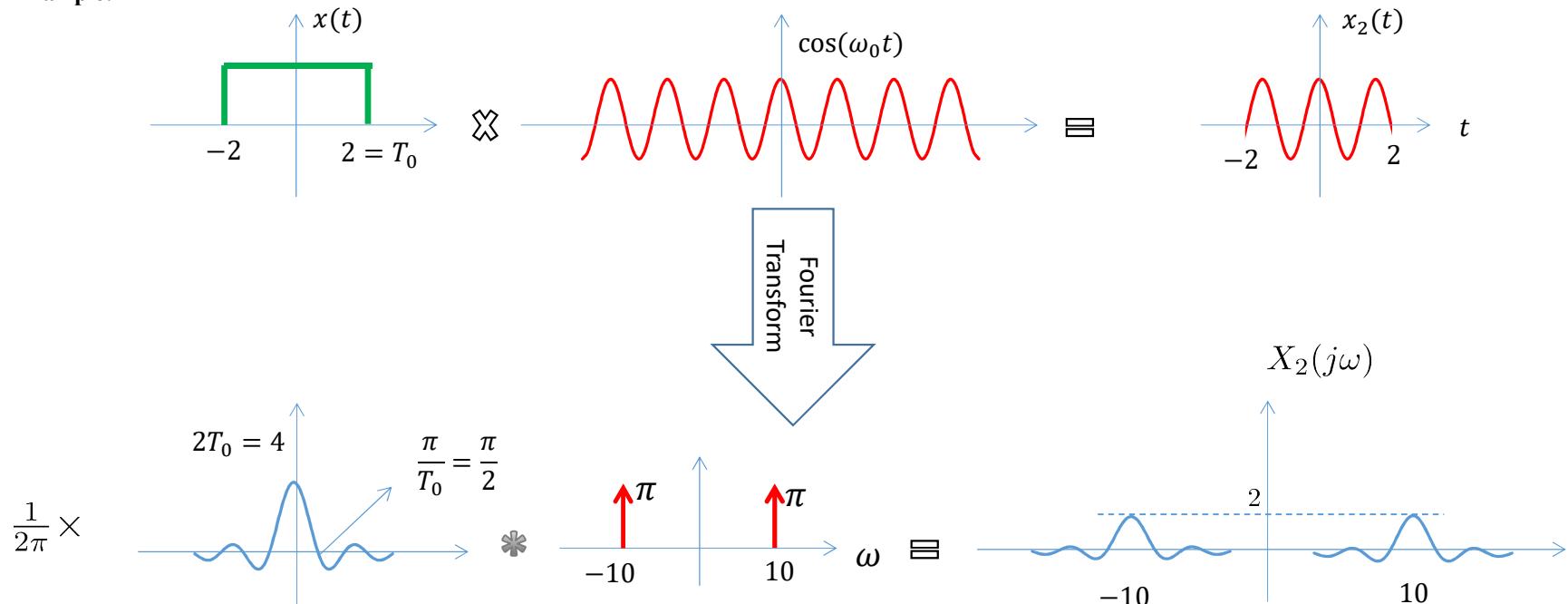
### 9. FT of product of signals

$$x_1(t) \xrightarrow{FT} X_1(j\omega)$$
$$x_2(t) \xrightarrow{FT} X_2(j\omega)$$
$$x_3(t) = x_1(t) \times x_2(t) \xrightarrow{FT} X_3(j\omega) = \frac{1}{2\pi} X_1(j\omega) * X_2(j\omega)$$

Product in time  $\equiv$  Convolution in Frequency

## Fourier Transform Properties (Product of Signals)

Example:

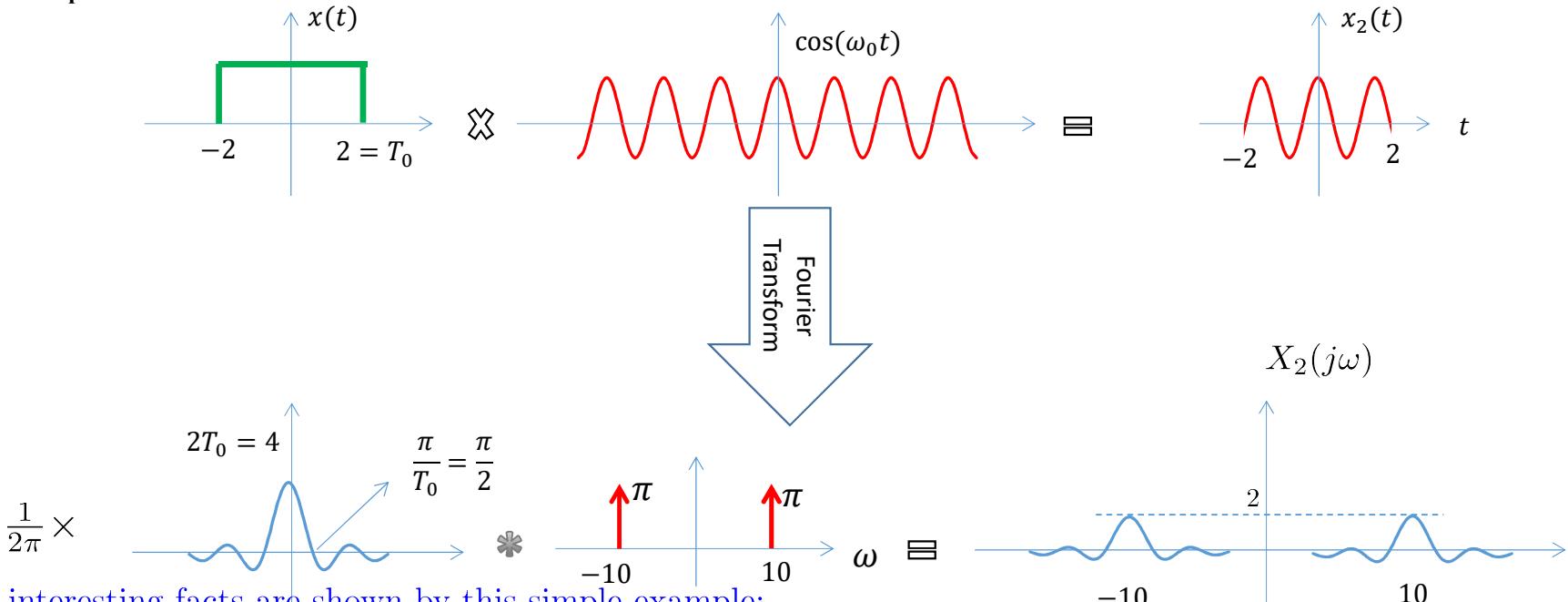


$$\begin{aligned}
 x_2(t) &= \cos(\omega_0 t) \times x(t), \quad \omega_0 = 10 \\
 x(t) \times \underbrace{\cos(10t + \theta)}_{\frac{1}{2}e^{j\theta}e^{j\omega_0 t} + \frac{1}{2}e^{-j\theta}e^{-j\omega_0 t}} &\xrightarrow{FT} \frac{1}{2\pi} X(j\omega) * (\pi e^{j\theta} \delta(j\omega - j\omega_0) + \pi e^{-j\theta} \delta(j\omega + j\omega_0)) \\
 &= \frac{1}{2} e^{j\theta} X(j(\omega - \omega_0)) + \frac{1}{2} e^{-j\theta} X(j(\omega + \omega_0))
 \end{aligned}$$

Windowed version of cosine

# Fourier Transform Properties (Product of Signals)

Example:



Two interesting facts are shown by this simple example:

- 1- **Multiplication** of a signal (here a cosine) with a **pulse windows** the signal in time and equivalently **convolves** the FT of the signal with **sinc function**.
- 2- **Multiplication** of a signal (here a pulse) with **cosine Amplitude Modulates(AM)** the signal in time and equivalently the FT of the signal **shifts** to  $\omega_o$  and  $-\omega_o$

## Fourier Transform Properties (Derivative and Integral)

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$
$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

### 9. FT of derivative of a signal

$$x(t) \xrightarrow{FT} X(j\omega)$$
$$v(t) = x'(t) \xrightarrow{FT} V(j\omega) = j\omega X(j\omega)$$

proof: take the derivative of both sides of the synthesis equation:

$$\frac{d}{dt} x(t) = \frac{d}{dt} \frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{d}{dt} X(j\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) \frac{d}{dt} e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) j\omega e^{j\omega t} d\omega$$

### 10. FT of integral of a signal

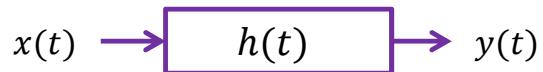
$$x(t) \xrightarrow{FT} X(j\omega)$$
$$z(t) = \int_{-\infty}^t x(\tau) d\tau \xrightarrow{FT} Z(j\omega) = \frac{X(j\omega)}{j\omega}$$

Note that the integral property can be used only if  $y(t) = \int_{-\infty}^t x(t) dt$  has a FT, i.e., the FT integral converges. For example it is known that  $FT(\delta(t)) = 1$  and  $u(t)$  is integral of  $\delta(t)$  but  $u(t)$  doesn't have a FT!

# Fourier Transform & Periodic Signals

FT properties	Signal	FT
	$x(t)$	$X(j\omega)$
	$z(t)$	$Z(j\omega)$
Linearity	$ax(t) + bz(t)$	$aX(j\omega) + bZ(j\omega)$
Time shift	$x(t - T_0)$	$e^{-j\omega T_0} X(j\omega)$
Freq. shift	$e^{j\omega_0 t} x(t)$	$X(j(\omega - \omega_0))$
Scaling	$x(at)$	$\frac{1}{ a } X(j\frac{\omega}{a})$
Duality	$X(jt)$	$2\pi x(-\omega)$
Complex Conj.	$x^*(t)$	$X^*(-j\omega)$ (so for real signals $ X(j\omega) $ is even and $\angle(X(j\omega))$ is odd)
Convolution	$x(t) * z(t)$	$X(j\omega) \times Z(j\omega)$
Periodic signals	$x_p(t)$ with $D_n$ coeffs	$X_p(j\omega) = \sum_n D_n 2\pi\delta(\omega - \omega_0 n)$
Product in time	$x(t) \times z(t)$	$\frac{1}{2\pi} X(j\omega) * Z(j\omega)$
Derivative	$x'(t)$	$j\omega X(j\omega)$
Integral	$\int_{-\infty}^t x(\tau) d\tau$	$\frac{X(j\omega)}{j\omega}$ only if the integral of FT converges

## Fourier Transform and LTI Systems



$$y(t) = h(t) * x(t), \quad Y(j\omega) = H(j\omega)X(j\omega),$$

Example:

$$H(j\omega) = \frac{1}{5 + 5j\omega}, \text{ system is causal}$$

Reminder:

$$X_1(j\omega) = e^{-j4\omega} \longrightarrow x_1(t) = \delta(t - 4)$$

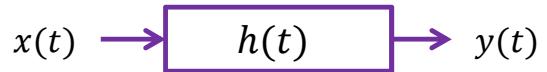
$$H(j\omega) = \frac{1}{5(j\omega + 1)} \longrightarrow h(t) = \frac{1}{5}e^{-t}u(t)$$

$$X_1(j\omega) = e^{-j4\omega} \quad \text{find} \quad y_1(t) = x_1(t) * h(t)$$

$$X_2(j\omega) = \frac{1}{2 + j\omega} \longrightarrow x_2(t) = e^{-2t}u(t)$$

$$X_2(j\omega) = \frac{1}{2 + j\omega} \quad \text{find} \quad y_2(t) = x_2(t) * h(t)$$

## Fourier Transform and LTI Systems



$$y(t) = h(t) * x(t), \quad Y(j\omega) = H(j\omega)X(j\omega),$$

$$H(j\omega) = \frac{1}{5 + 5j\omega}$$

Method 1:

Find convolutions directly:

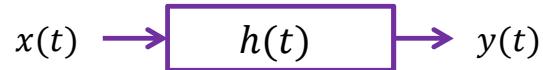
$$X_1(j\omega) = e^{-j4\omega} \quad \text{find} \quad y_1(t) = x_1(t) * h(t)$$

$$y_1(t) = \delta(t - 4) * \left( \frac{1}{5}e^{-t}u(t) \right)$$

$$X_2(j\omega) = \frac{1}{2 + j\omega} \quad \text{find} \quad y_2(t) = x_2(t) * h(t)$$

$$y_2(t) = e^{-2t}u(t) * \left( \frac{1}{5}e^{-t}u(t) \right)$$

## Fourier Transform and LTI Systems



$$H(j\omega) = \frac{1}{5 + 5j\omega}$$

$$X_1(j\omega) = e^{-j4\omega} \text{ find } y_1(t) = x_1(t) * h(t)$$

$$X_2(j\omega) = \frac{1}{2 + j\omega} \text{ find } y_2(t) = x_2(t) * h(t)$$

$$y(t) = h(t) * x(t), \quad Y(j\omega) = H(j\omega)X(j\omega),$$

Method 2:

Find  $Y_1(j\omega)$  &  $Y_2(j\omega)$  first and then take IFT of them to find  $y_1(t)$  &  $y_2(t)$ .

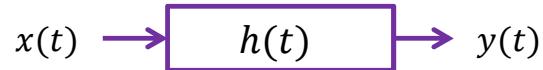
$$Y_1(j\omega) = X_1(j\omega)H(j\omega) = e^{-j4\omega} \frac{1}{5 + 5j\omega} \xrightarrow{IFT} \delta(t - 4) * h(t) = \frac{1}{5} e^{-(t-4)} u(t - 4)$$

$$Y_2(j\omega) = X_2(j\omega)H(j\omega) = \frac{1}{2 + j\omega} \times \frac{1}{5(1 + j\omega)}$$

Use Partial Fraction Expansion and fine  $a$  and  $b$  to rewrite  $Y_2(j\omega)$  as

$$Y_2(j\omega) = \frac{a}{2 + j\omega} + \frac{b}{1 + j\omega}$$

# Fourier Transform and LTI Systems



$$H(j\omega) = \frac{1}{5 + 5j\omega}$$

$$X_1(j\omega) = e^{-j4\omega} \text{ find } y_1(t) = x_1(t) * h(t)$$

$$X_2(j\omega) = \frac{1}{2 + j\omega} \text{ find } y_2(t) = x_2(t) * h(t)$$

$$y(t) = h(t) * x(t), \quad Y(j\omega) = H(j\omega)X(j\omega),$$

Method 2:

Find  $Y_1(j\omega)$  &  $Y_2(j\omega)$  first and then take IFT of them to find  $y_1(t)$  &  $y_2(t)$ .

$$Y_1(j\omega) = X_1(j\omega)H(j\omega) = e^{-j4\omega} \frac{1}{5 + 5j\omega} \xrightarrow{IFT} \delta(t - 4) * h(t) = \frac{1}{5} e^{-(t-4)} u(t - 4)$$

$$Y_2(j\omega) = X_2(j\omega)H(j\omega) = \frac{1}{2 + j\omega} \times \frac{1}{5(1 + j\omega)}$$

$$Y_2(j\omega) = \frac{\frac{1}{5}}{1 + j\omega} + \frac{-\frac{1}{5}}{2 + j\omega}$$

Use Partial Fraction Expansion and fine  $a$  and  $b$  to rewrite  $Y_2(j\omega)$  as

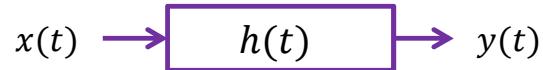
$$Y_2(j\omega) = \frac{a}{2 + j\omega} + \frac{b}{1 + j\omega}$$

$$a = (2 + j\omega)Y_2(j\omega)|_{j\omega=-2} = -\frac{1}{5}$$

$$b = (1 + j\omega)Y_2(j\omega)|_{j\omega=-1} = \frac{1}{5}$$

↓  
IFT

# Fourier Transform and LTI Systems



$$H(j\omega) = \frac{1}{5 + 5j\omega}$$

$$X_1(j\omega) = e^{-j4\omega} \text{ find } y_1(t) = x_1(t) * h(t)$$

$$X_2(j\omega) = \frac{1}{2 + j\omega} \text{ find } y_2(t) = x_2(t) * h(t)$$

$$y(t) = h(t) * x(t), \quad Y(j\omega) = H(j\omega)X(j\omega),$$

Method 2:

Find  $Y_1(j\omega)$  &  $Y_2(j\omega)$  first and then take IFT of them to find  $y_1(t)$  &  $y_2(t)$ .

$$Y_1(j\omega) = X_1(j\omega)H(j\omega) = e^{-j4\omega} \frac{1}{5 + 5j\omega} \xrightarrow{IFT} \delta(t - 4) * h(t) = \frac{1}{5} e^{-(t-4)} u(t - 4)$$

$$Y_2(j\omega) = X_2(j\omega)H(j\omega) = \frac{1}{2 + j\omega} \times \frac{1}{5(1 + j\omega)}$$

$$Y_2(j\omega) = \frac{\frac{1}{5}}{1 + j\omega} + \frac{-\frac{1}{5}}{2 + j\omega}$$

Use Partial Fraction Expansion and fine  $a$  and  $b$  to rewrite  $Y_2(j\omega)$  as

$$Y_2(j\omega) = \frac{a}{2 + j\omega} + \frac{b}{1 + j\omega}$$

$$a = (2 + j\omega)Y_2(j\omega)|_{j\omega=-2} = -\frac{1}{5}$$

$$b = (1 + j\omega)Y_2(j\omega)|_{j\omega=-1} = \frac{1}{5}$$

$$y_2(t) = \frac{1}{5}e^{-t}u(t) - \frac{1}{5}e^{-2t}u(t)$$

↓  
IFT

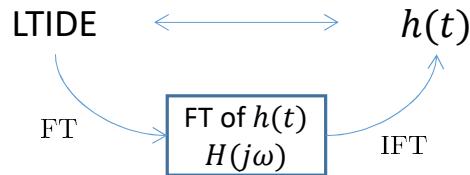
Useful properties:

$$\begin{aligned} e^{j\omega_0 t} X(j\omega) &\xrightarrow{\text{IFT}} x(t - t_0) \\ j\omega X(j\omega) &\xrightarrow{\text{IFT}} x'(t) \\ (j\omega)^2 X(j\omega) &\xrightarrow{\text{IFT}} x''(t) \end{aligned}$$

## Fourier Transform and LTIDE Systems

Find  $h(t)$  for the following system:

$$y''(t) + ay'(t) + by(t) = cx'(t) + dx(t)$$



New approach:

$$FT(y''(t) + ay'(t) + by(t)) = FT(cx'(t) + dx(t))$$

$$(j\omega)^2 Y(j\omega) + a(j\omega)Y(j\omega) + bY(j\omega) = c(j\omega)X(j\omega) + dX(j\omega)$$

$$Y(j\omega)((j\omega)^2 + a(j\omega) + b) = X(j\omega)(c(j\omega) + d)$$

$$Y(j\omega) = X(j\omega) \frac{c(j\omega) + d}{(j\omega)^2 + a(j\omega) + b}$$

$$H(j\omega) = \frac{Y(j\omega)}{X(j\omega)} = \frac{c(j\omega) + d}{(j\omega)^2 + a(j\omega) + b} \quad \xrightarrow{\text{IFT}} \quad h(t)$$

$h(t)$  is response to  $x(t) = \delta(t)$  or equivalently is found by setting  $X(j\omega) = 1$ .

## Fourier Transform and LTIDE Systems

**Example:** Find  $h(t)$  for the following causal system by first finding  $H(j\omega)$

$$y''(t) + 4y'(t) + 4y(t) = 2x'(t) + 2x(t)$$

## Fourier Transform and LTIDE Systems

**Example:** Find  $h(t)$  for the following causal system by first finding  $H(j\omega)$

$$y''(t) + 4y'(t) + 4y(t) = 2x'(t) + 2x(t)$$

Solution:

$$((j\omega)^2 + 4j\omega + 4) Y(j\omega) = 2(j\omega + 1)X(j\omega)$$

$$Y(j\omega) = \underbrace{\frac{2(j\omega + 1)}{(j\omega)^2 + 4j\omega + 4}}_{H(j\omega)} X(j\omega)$$

$$H(j\omega) = \frac{2(j\omega + 1)}{(j\omega)^2 + 4j\omega + 4}$$

Two methods can be followed to find  $h(t)$

# Fourier Transform and LTIDE Systems

Method 1: Use Partial fractional expansion (PFE)

$$H(j\omega) = \frac{2(j\omega+1)}{(j\omega)^2 + 4j\omega + 4}$$
$$H(j\omega) = \frac{A}{j\omega + 2} + \frac{B}{(j\omega + 2)^2}$$
$$B = H(j\omega)(j\omega + 2)^2 \Big|_{j\omega=-2} = 2(j\omega + 1) \Big|_{j\omega=-2} = -2$$
$$A = \frac{d}{dj\omega} (H(j\omega)(j\omega + 2)^2) \Big|_{j\omega=-2} = \frac{d}{dj\omega} 2(j\omega + 1) = 2$$
$$H(j\omega) = \frac{2}{j\omega + 2} + \frac{-2}{(j\omega + 2)^2}$$

$$\frac{2}{j\omega + 2} \xrightarrow{IFT} 2e^{-2t}u(t)$$

$$\frac{-2}{(j\omega + 2)^2} \xrightarrow{IFT} -2te^{-2t}u(t)$$

$$\text{Reminder: } \frac{1}{(j\omega + a)^2} \xrightarrow{IFT} te^{-at}u(t)$$

$$h(t) = (2e^{-2t} - 2te^{-2t})u(t)$$

# Fourier Transform and LTIDE Systems

$$H(j\omega) = \frac{2(j\omega+1)}{(j\omega)^2 + 4j\omega + 4}$$

Method 2:

$$\begin{aligned} H(j\omega) &= \frac{2 + 2j\omega}{(j\omega + 2)^2} = \frac{2}{(j\omega + 2)^2} + \frac{2j\omega}{(j\omega + 2)^2} \\ \frac{2}{(j\omega + 2)^2} &\xrightarrow{IFT} 2te^{-2t}u(t) \\ j\omega \cdot \frac{2}{(j\omega + 2)^2} &\xrightarrow{IFT} \frac{d}{dt}(2te^{-2t}u(t)) = 2e^{-2t}u(t) - 4te^{-2t}u(t) \end{aligned}$$

$$\begin{aligned} h(t) &= 2te^{-2t}u(t) + 2e^{-2t}u(t) - 4te^{-2t}u(t) \\ &= 2e^{-2t}u(t) - 2te^{-2t}u(t) \end{aligned}$$

## Fourier Transform and LTIDE Systems

**Example:** Find the impulse response of the following causal system and also its output to  $x(t) = e^{-6t}u(t)$  by using FT and  $H(j\omega)$ .

$$3y(t) + \frac{dy}{dt} = 8x(t) + 2\frac{dx}{dt}$$

# Fourier Transform and LTIDE Systems

**Example:** Find the impulse response of the following causal system and also its output to  $x(t) = e^{-6t}u(t)$  by using FT and  $H(j\omega)$ .

$$3y(t) + \frac{dy}{dt} = 8x(t) + 2\frac{dx}{dt}$$

Reminder: In this example  $M = N$  and we have  $b_0\delta(t)$  term in  $h(t)$

$$H(j\omega) = \frac{8 + 2j\omega}{3 + j\omega} = 2 + \frac{2}{3 + j\omega}$$

$$Y(j\omega) = \frac{8 + 2j\omega}{3 + j\omega} \times \frac{1}{6 + j\omega} = \frac{a}{3 + j\omega} + \frac{b}{6 + j\omega}$$

$$h(t) = 2\delta(t) + 2e^{-3t}u(t)$$

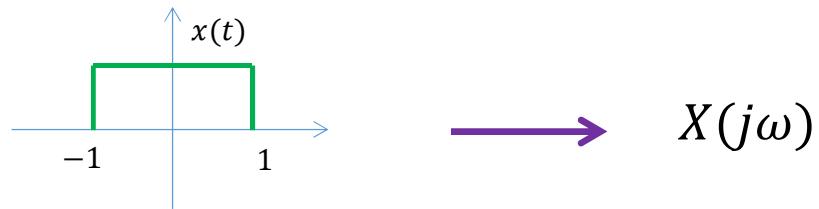
$$y(t) = \frac{2}{3}e^{-3t}u(t) + \frac{4}{3}e^{-6t}u(t)$$

Note: In using partial fraction expansion order of  $j\omega$  in the numerator has to be less than the order of  $j\omega$  in the denominator.

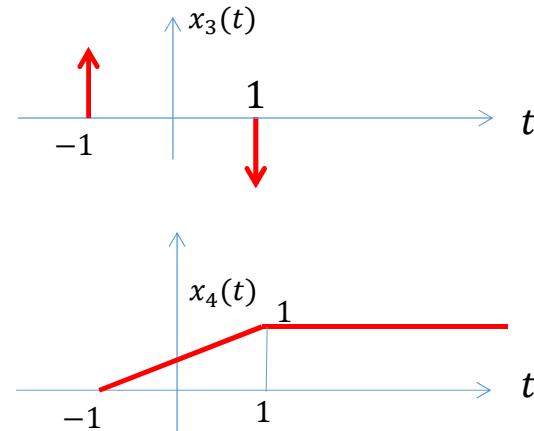
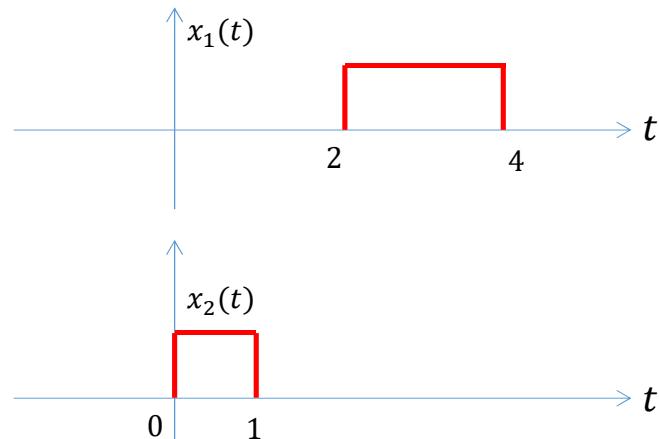
This method can be used for cases where  $M$  (order of highest derivate of input) is even greater than  $N$  (system order that is the order of highest derivative of output).

## Using FT Properties

Consider the following signal

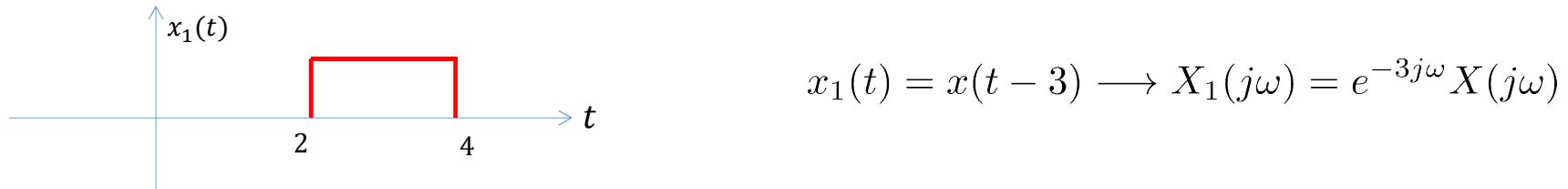
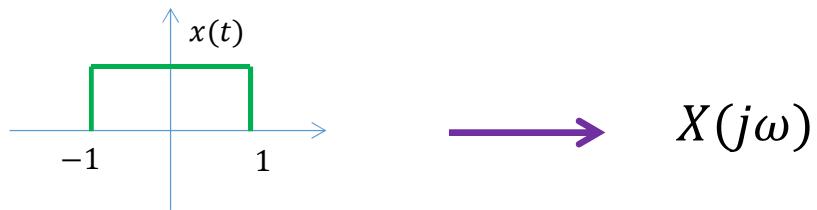


Write the FT of the following signals as a function of  $X(j\omega)$ . ( Note this is not asking for finding the FT of the signals directly). Your answer must be only function of  $X(j\omega)$ .

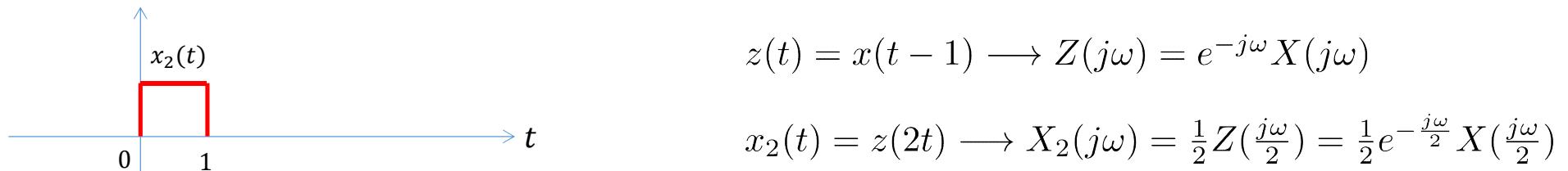


## Using FT Properties

Consider the following signal

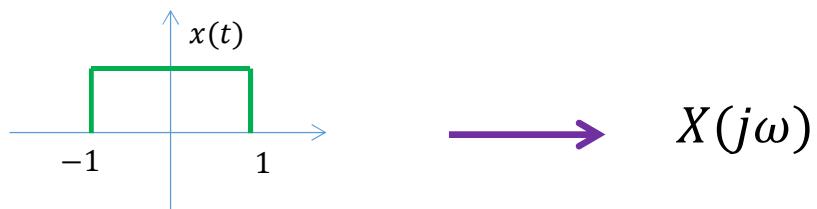


$$x_2(t) = x(2t - 1)$$

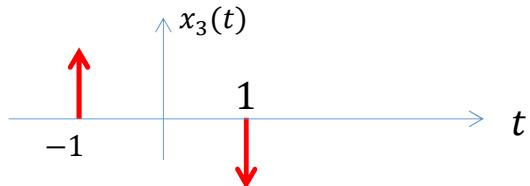


## Using FT Properties

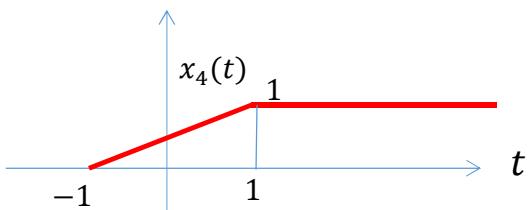
Consider the following signal



$$\longrightarrow X(j\omega)$$



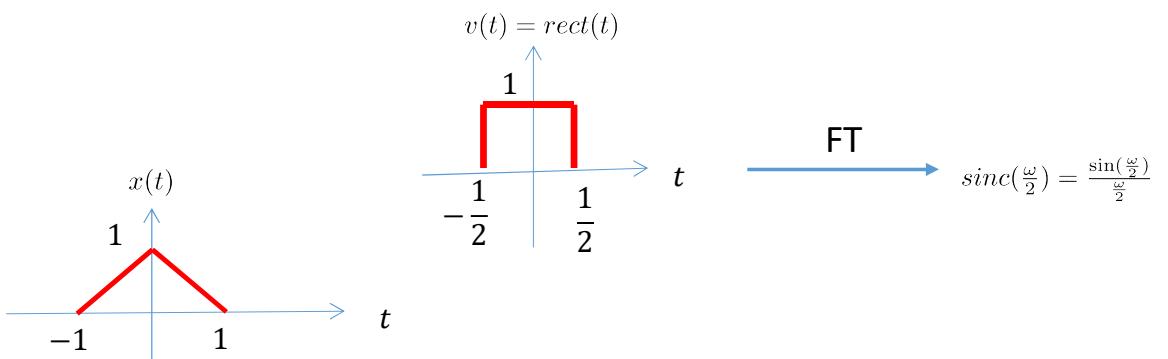
$$x_3(t) = \frac{d}{dt}x(t) \longrightarrow X_3(j\omega) = j\omega X(j\omega)$$



$$x_4(t) = \int_{-\infty}^t x(t)dt \longrightarrow X_4(j\omega) = \frac{1}{j\omega}X(j\omega)$$

## Using FT Properties

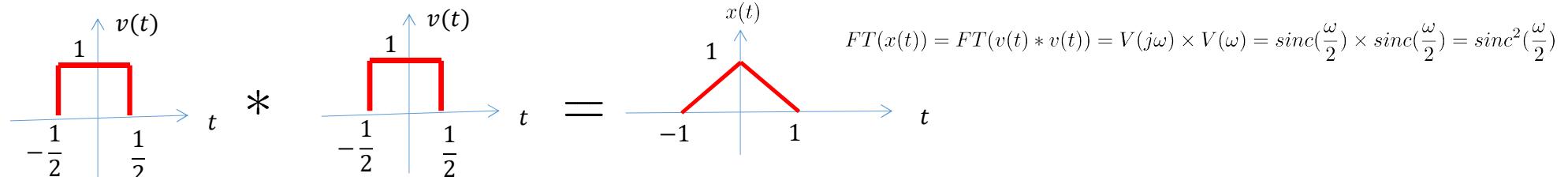
Find Fourier Transform of  $x(t)$ :



Method 1: Use the integral

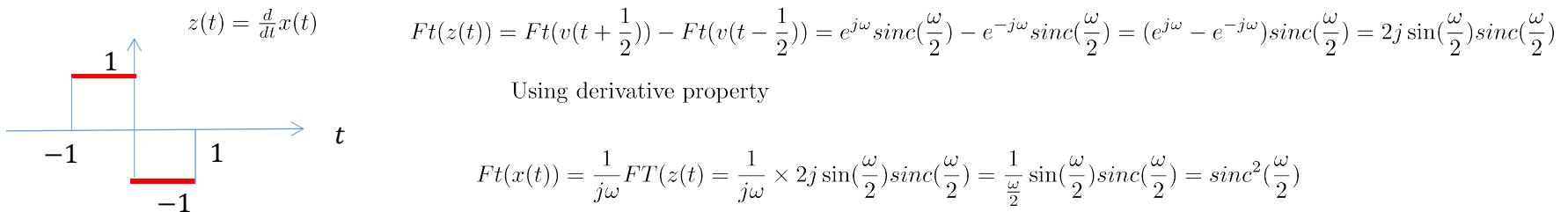
$$X(j\omega) = \int_{-1}^1 e^{-j\omega t} x(t) dt = \int_{-1}^0 (t+1)e^{-j\omega t} dt + \int_0^1 (-t+1)e^{-j\omega t} dt.$$

Method 2: Use FT properties Table:



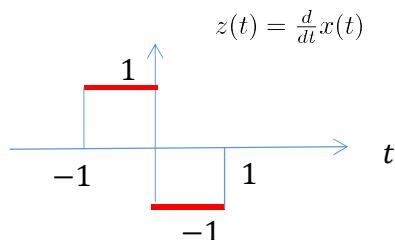
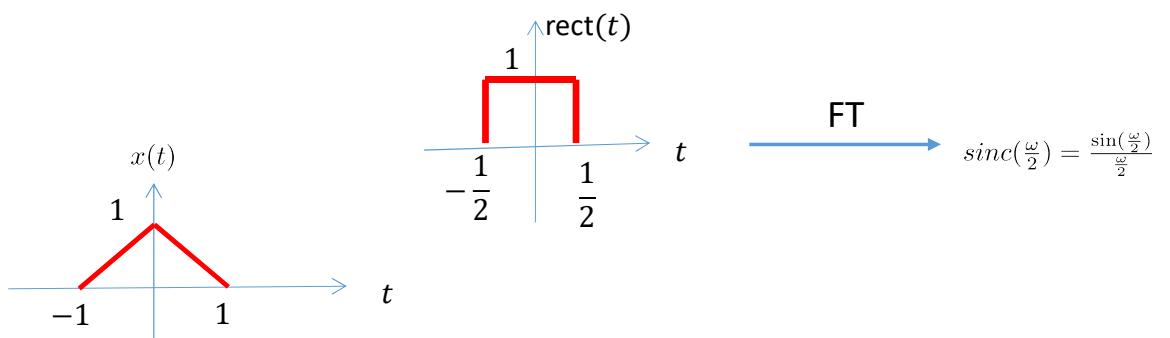
Or:

Using Linearity and shift properties and then Euler's formula:



## Using FT Properties

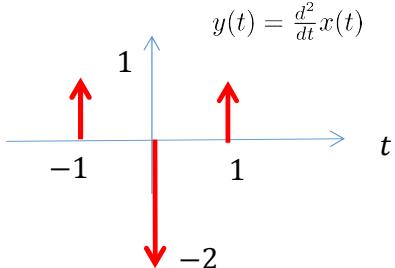
Find Fourier Transform of  $x(t)$  by using its second derivatives:



$$Ft(y(t)) = e^{j\omega} - 2 + e^{-j\omega} = 2 \cos(j\omega) - 2$$

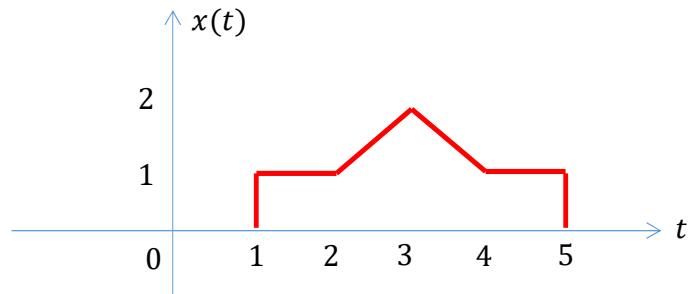
$$Ft(x(t)) = \frac{1}{j\omega} Ft(z(t)) = \frac{1}{j\omega} \times \frac{1}{j\omega} Ft(y(t)) = \frac{1}{-\omega^2} Ft(y(t))$$

$$= \frac{2(1 - \cos(\omega))}{\omega^2} = \frac{4 \sin^2(\frac{\omega}{2})}{\omega^2} = \text{sinc}^2(\frac{\omega}{2})$$



## Using FT Properties

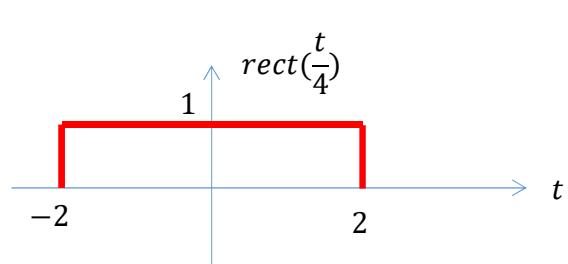
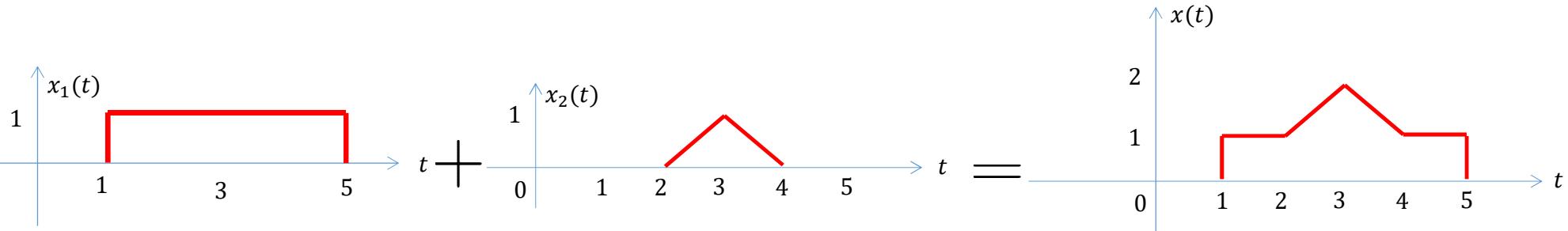
**Example:** Find the FT of the following signal:



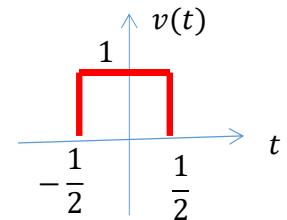
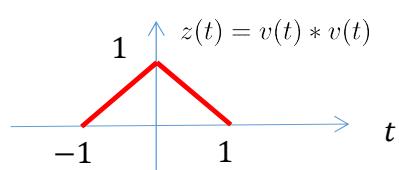
Method 1: Use the difficult integral  $X(j\omega) = \int_1^5 e^{-j\omega t} x(t) dt$ .

Method 2: Use FT properties Table

## Using FT Properties



$$\begin{aligned}FT(x_1(t)) &= FT(\text{rect}(\frac{t-3}{4})) \\&= e^{-3j\omega} FT(\text{rect}(\frac{t}{4})) \\&= e^{-3j\omega} 4 \text{sinc}(\frac{4\omega}{2})\end{aligned}$$



$$x(t) = x_1(t) + x_2(t)$$

$$FT(x_2(t)) = FT(z(t-3)) = e^{-3j\omega} FT(z(t))$$

$$FT(z(t)) = FT(v(t) * v(t)) = V(j\omega) \times V(j\omega) = \text{sinc}^2(\frac{\omega}{2})$$

$$FT(x_2(t)) = e^{-3j\omega} \text{sinc}^2(\frac{\omega}{2})$$

$$FT(x(t)) = FT(x_1(t) + x_2(t)) = e^{-3j\omega} 4 \text{sinc}(\frac{4\omega}{2}) + e^{-3j\omega} \text{sinc}^2(\frac{\omega}{2})$$

# Using FT Properties

Or use derivative of  $x(t)$  to find its FT:

$$z(t) = \delta(t - 1) + v(t - \frac{5}{2}) - v(t - \frac{7}{2}) - \delta(t - 5)$$

$$FT(z(t)) = e^{-j\omega} + e^{-\frac{5}{2}j\omega} FT(v(t)) - e^{-\frac{7}{2}j\omega} FT(v(t)) - e^{-j5\omega}$$

$$= e^{-j\omega} - e^{-j5\omega} + e^{-3j\omega}(e^{\frac{1}{2}j\omega} - e^{-\frac{1}{2}j\omega}) FT(v(t))$$

$$= e^{-j3\omega}(e^{j2\omega} - e^{-j2\omega}) +$$

$$= 2je^{-3j\omega} \sin(2\omega) + 2je^{-3j\omega} \sin(\frac{\omega}{2}) FT(v(t))$$

$$= 2je^{-j3\omega}(\sin(2\omega) + \sin(\frac{\omega}{2}) \text{sinc}(\frac{\omega}{2}))$$

$$FT(x(t)) = \frac{1}{j\omega} FT(z(t))$$

$$= e^{-j3\omega}(\frac{4}{2\omega} \sin(2\omega) + \frac{1}{\omega} \sin(\frac{\omega}{2}) \text{sinc}(\frac{\omega}{2}))$$

$$= e^{-j3\omega}(4\text{sinc}(2\omega) + \text{sinc}^2(\frac{\omega}{2}))$$

alternatively we can first work on shifted version of  $z(t)$  that is symmetric. Compare the steps:

$$y(t) = z(t + 3)$$

$$y(t) = \delta(t + 2) + v(t + \frac{1}{2}) - v(t - \frac{1}{2}) - \delta(t - 2)$$

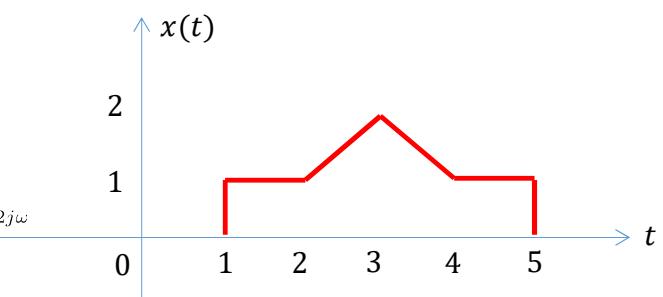
$$FT(y(t)) = e^{2j\omega} + e^{+\frac{1}{2}j\omega} FT(v(t)) - e^{-\frac{1}{2}j\omega} FT(v(t)) - e^{-2j\omega}$$

$$= e^{2j\omega} - e^{-2j\omega} + (e^{\frac{1}{2}j\omega} - e^{-\frac{1}{2}j\omega}) FT(v(t))$$

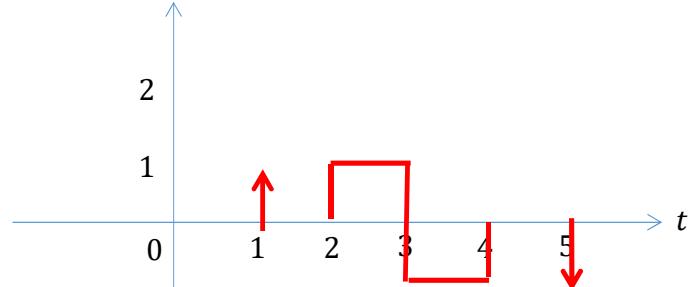
$$= 2j \sin(2\omega) + \sin(\frac{\omega}{2}) FT(v(t))$$

$$= 2j \sin(2\omega) + \sin(\frac{\omega}{2}) \text{sinc}(\frac{\omega}{2})$$

$$FT(z(t)) = e^{-j3\omega} FT(y(t))$$



$$z(t) = \frac{d}{dt} x(t)$$



## Reminder: Fourier Transform Properties (Frequency Shift)

Fourier Transform of causal part of real part of an exponential signal ( $\sigma > 0$ ):

$$x(t) = e^{-\sigma t} \cos(\omega_0 t) u(t)$$

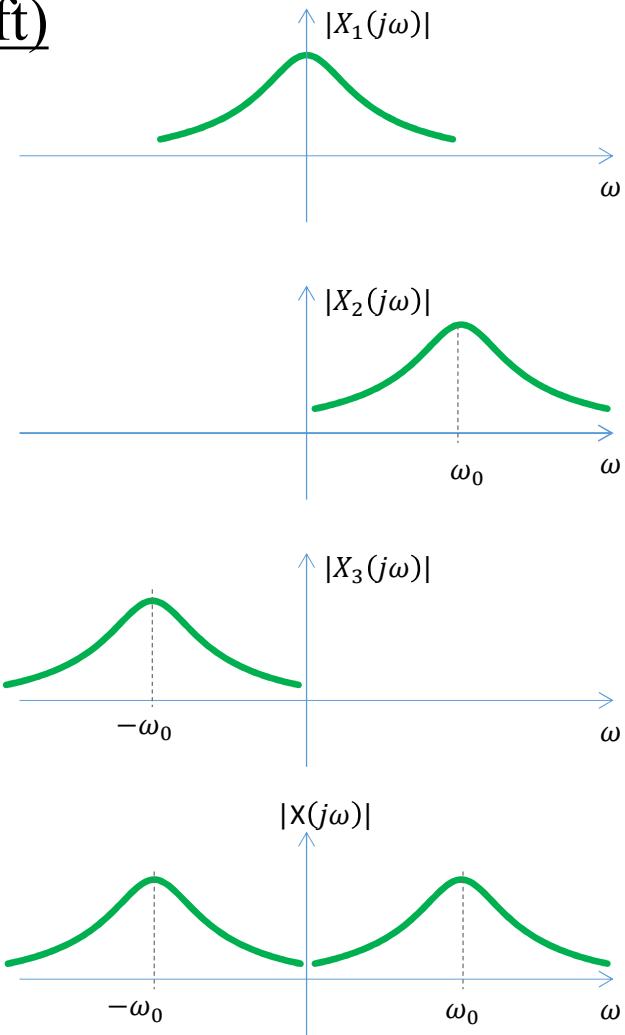
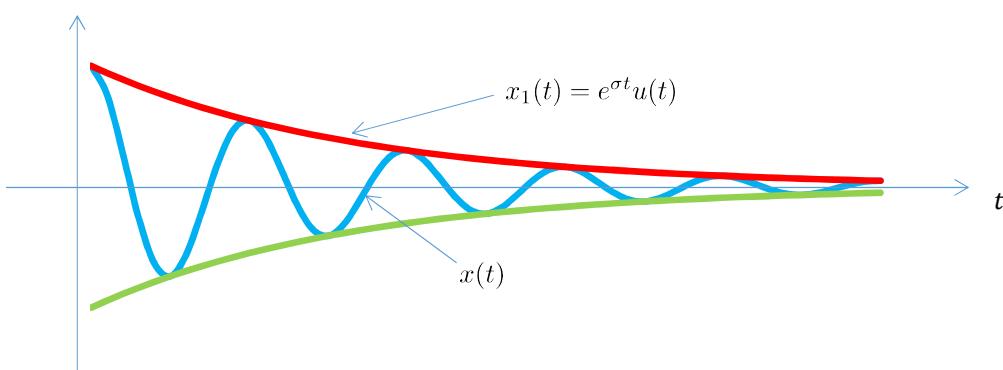
$$x_1(t) = e^{-\sigma t} u(t) \quad FT(x_1(t)) = X_1(j\omega) = \frac{1}{j\omega + \sigma}$$

$$x(t) = \frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2} x_1(t)$$

$$FT(x(t)) = \frac{1}{2} (X_1(j\omega - \omega_0) + X_1(j\omega + \omega_0))$$

$$= \frac{1}{2} \left( \frac{1}{j(\omega - \omega_0) + \sigma} + \frac{1}{j(\omega + \omega_0) + \sigma} \right) = \frac{1}{2} \left( \frac{1}{j\omega + \sigma - j\omega_0} + \frac{1}{j\omega + \sigma + j\omega_0} \right)$$

$$= \frac{1}{2} \frac{j\omega + \sigma - j\omega_0 + j\omega + \sigma + j\omega_0}{(j\omega + \sigma)^2 - (j\omega_0)^2} = \frac{j\omega + \sigma}{(j\omega + \sigma)^2 + \omega_0^2}$$



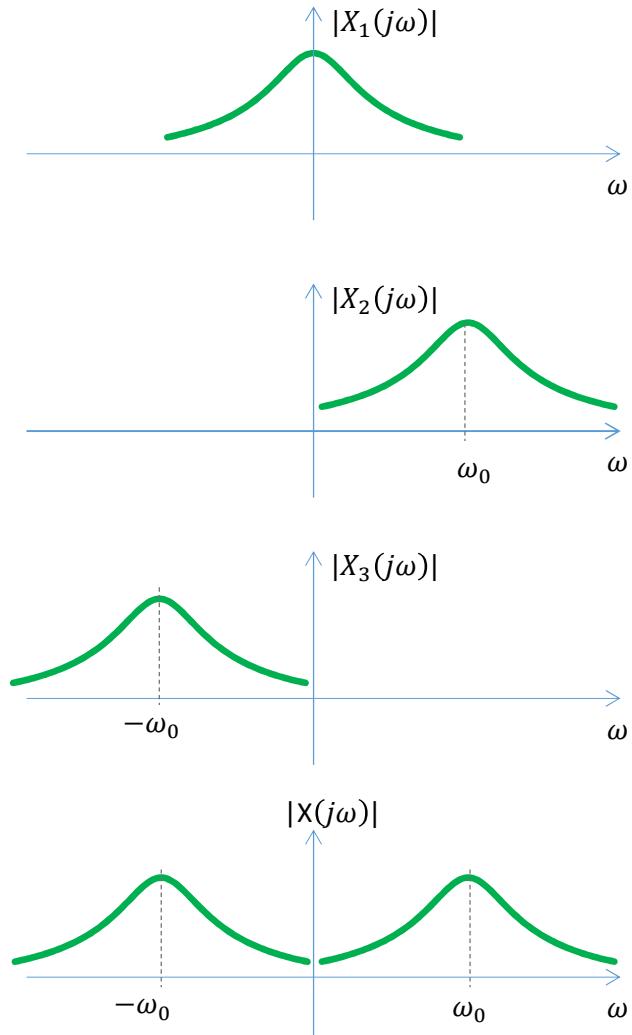
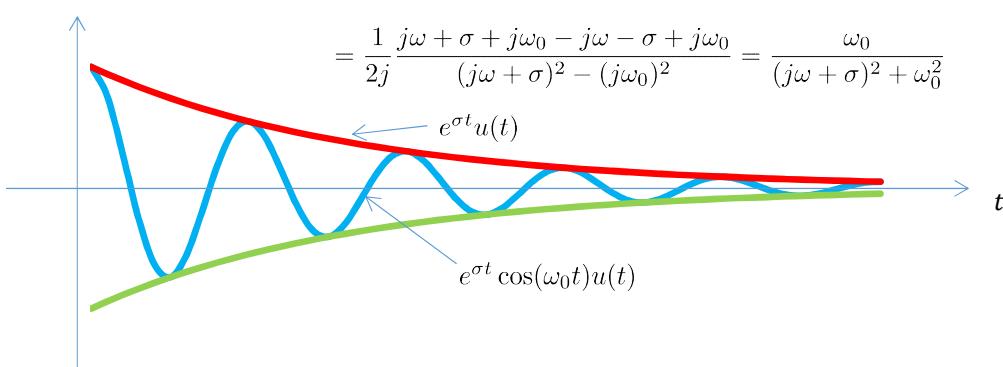
# Using Fourier Transform Properties

Fourier Transform of causal part of an exponential signal  $\sigma > 0$ :

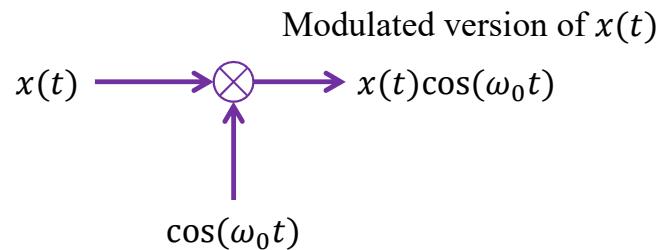
$$x(t) = e^{-st}u(t) = e^{-(\sigma+j\omega_0)t}u(t) \quad X(j\omega) = \frac{1}{s+j\omega}$$

$$\begin{aligned} FT(e^{-\sigma t} \cos(\omega_0 t)u(t)) &= \frac{1}{2}(FT(e^{(-\sigma+j\omega_0)t}u(t) + FT(e^{(-\sigma-j\omega_0)t}u(t))) \\ &= \frac{1}{2}\left(\frac{1}{\sigma-j\omega_0+j\omega} + \frac{1}{\sigma+j\omega_0+j\omega}\right) \\ &= \frac{1}{2} \frac{j\omega + \sigma + j\omega_0 + j\omega + \sigma - j\omega_0}{(j\omega + \sigma)^2 - (j\omega_0)^2} = \frac{j\omega + \sigma}{(j\omega + \sigma)^2 + \omega_0^2} \end{aligned}$$

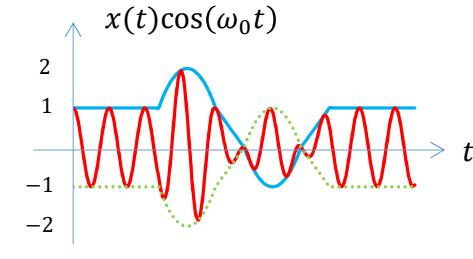
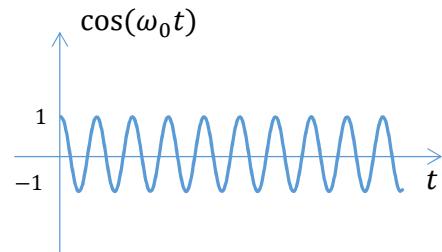
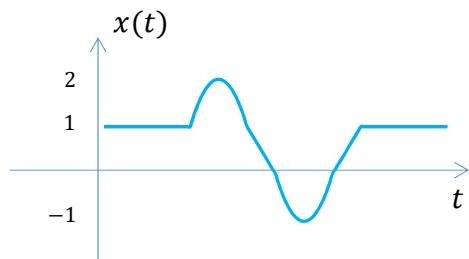
$$\begin{aligned} FT(e^{-\sigma t} \sin(\omega_0 t)u(t)) &= \frac{1}{2j}(FT(e^{(-\sigma+j\omega_0)t}u(t) - FT(e^{(-\sigma-j\omega_0)t}u(t))) \\ &= \frac{1}{2j}\left(\frac{1}{\sigma-j\omega_0+j\omega} - \frac{1}{\sigma+j\omega_0+j\omega}\right) \\ &= \frac{1}{2j} \frac{j\omega + \sigma + j\omega_0 - j\omega - \sigma + j\omega_0}{(j\omega + \sigma)^2 - (j\omega_0)^2} = \frac{\omega_0}{(j\omega + \sigma)^2 + \omega_0^2} \end{aligned}$$



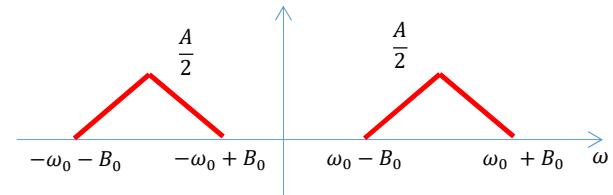
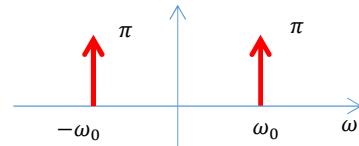
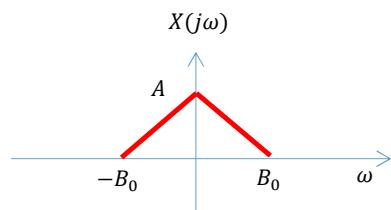
# Modulation



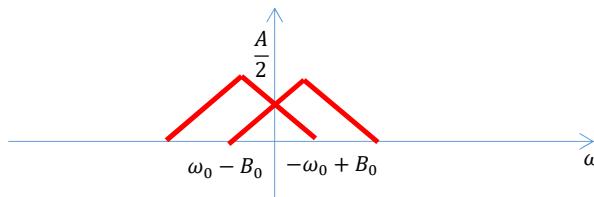
$$x_1(t)x_2(t) \xrightarrow{FT} \frac{1}{2\pi} X_1(j\omega) * X_2(j\omega)$$



$$\frac{1}{2\pi} X(j\omega) * (\pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0))$$

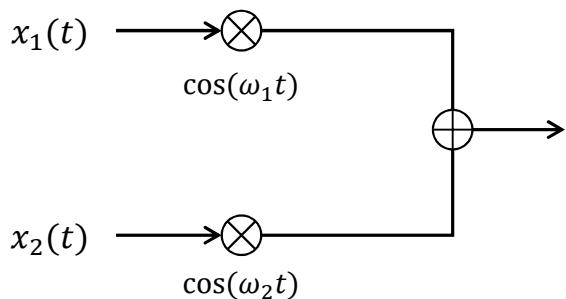


If  $\omega_0 < B_0$   
Aliasing

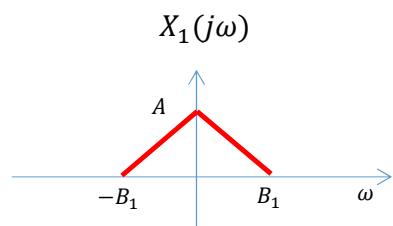
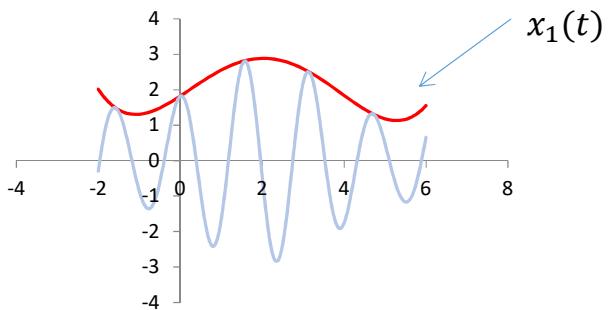
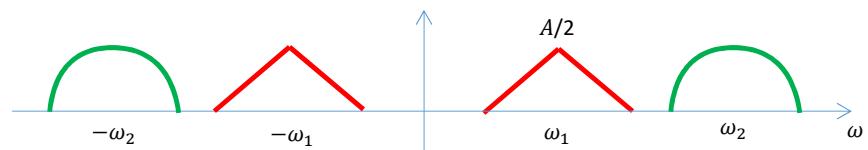


# Modulation

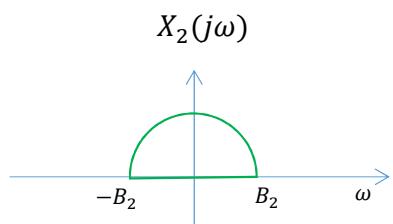
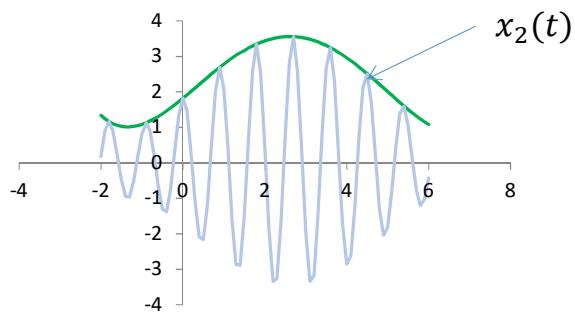
Two signals over the same channel



$$FT(x_1(t) \cos(\omega_1 t) + x_2(t) \cos(\omega_2 t))$$



$$FT(\cos(\omega_1 t))$$



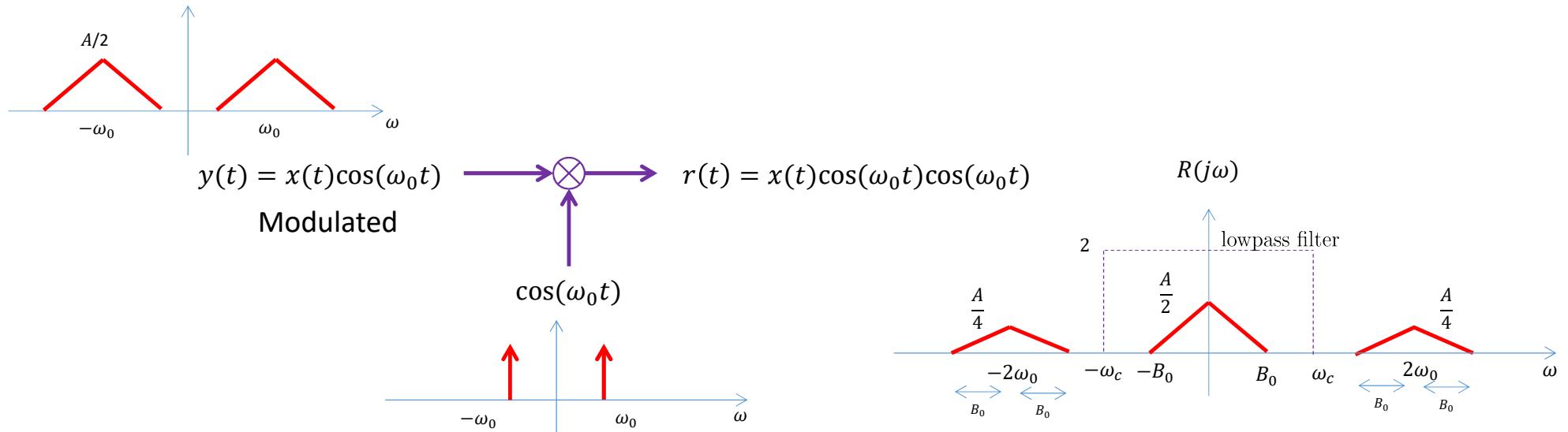
$$FT(\cos(\omega_2 t))$$



What are the conditions so that shapes of  $X_1(j\omega)$  &  $X_2(j\omega)$  are preserved?

# Demodulation

At the receiver multiply the signal to the same  $\cos(\omega_0 t)$



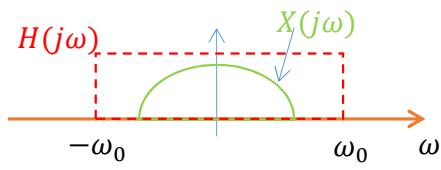
$$r(t) = x(t) \cos^2(\omega_0 t) = x(t) \left( \frac{1 + \cos(2\omega_0 t)}{2} \right) = \frac{1}{2}x(t) + \frac{1}{2} \cos(2\omega_0 t)x(t)$$

$$x_1(t)x_2(t) \xrightarrow{FT} \frac{1}{2\pi} X_1(j\omega) * X_2(j\omega)$$

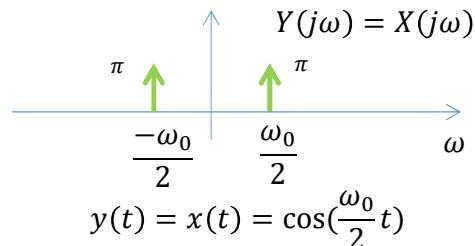
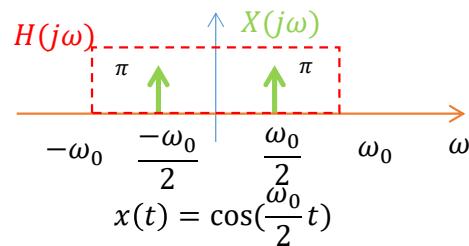
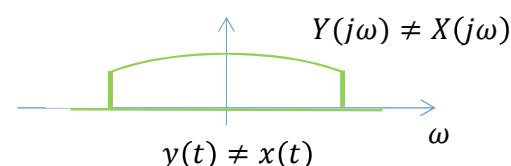
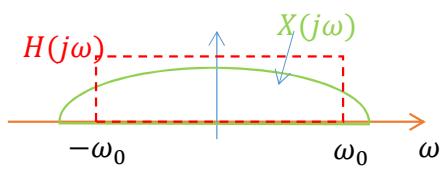
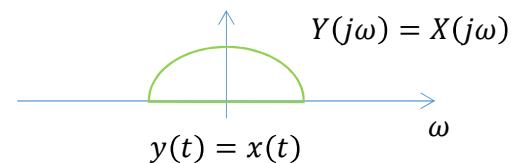
$$R(j\omega) = \frac{1}{2}X(j\omega) + \frac{1}{4}(X(j\omega - 2\omega_0) + X(j\omega + 2\omega_0))$$

## Lowpass Filter and signal recovery

$$y(t) = x(t) * h(t)$$

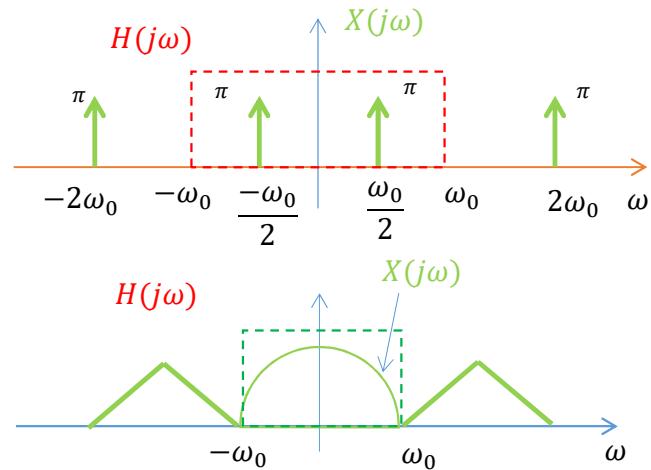


$$Y(j\omega) = X(j\omega)H(j\omega)$$

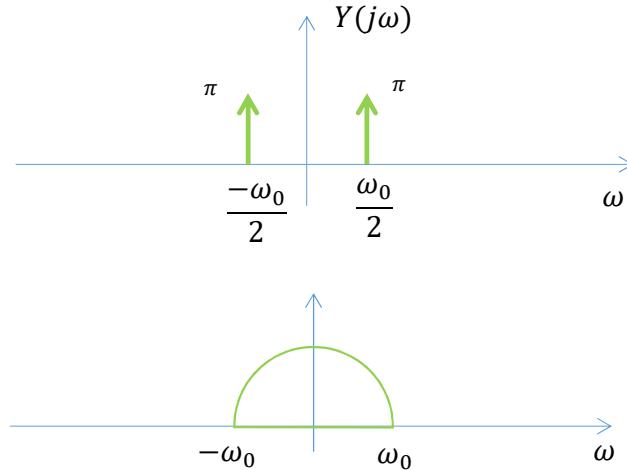


# Demodulation

$$x(t) = \cos\left(\frac{\omega_0}{2} t\right) + \cos(2\omega_0)$$

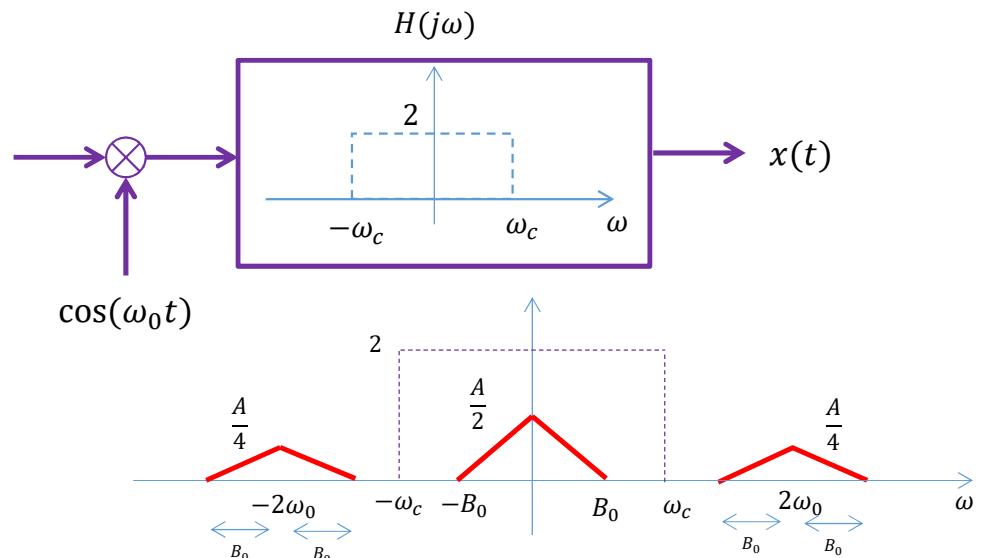
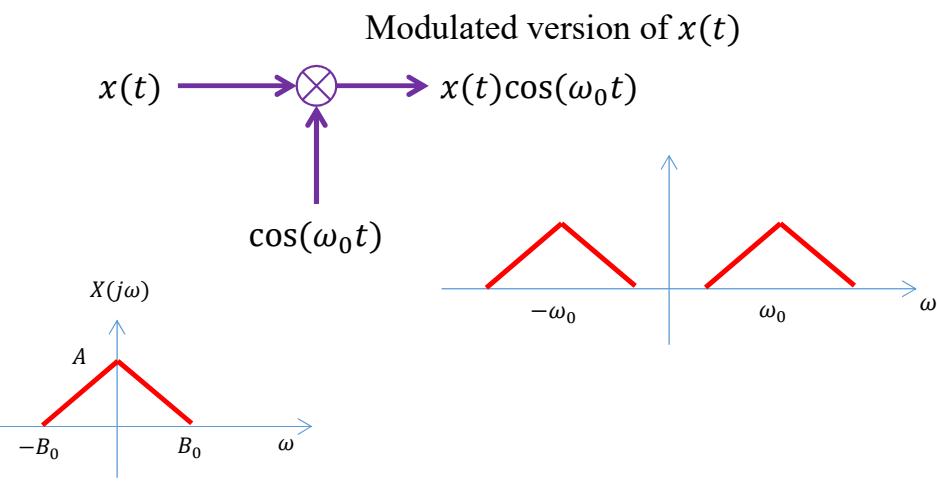


$$y(t) = \cos\left(\frac{\omega_0}{2} t\right)$$



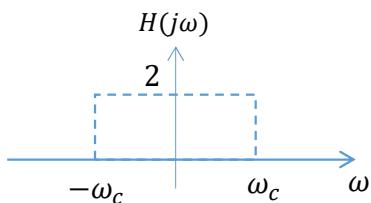
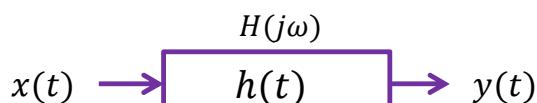
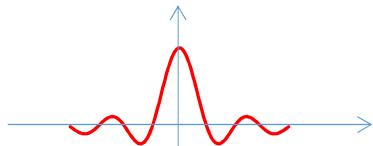
# Modulation and Demodulation

Next pass the received signal through a low pass filter!



Low-pass filter

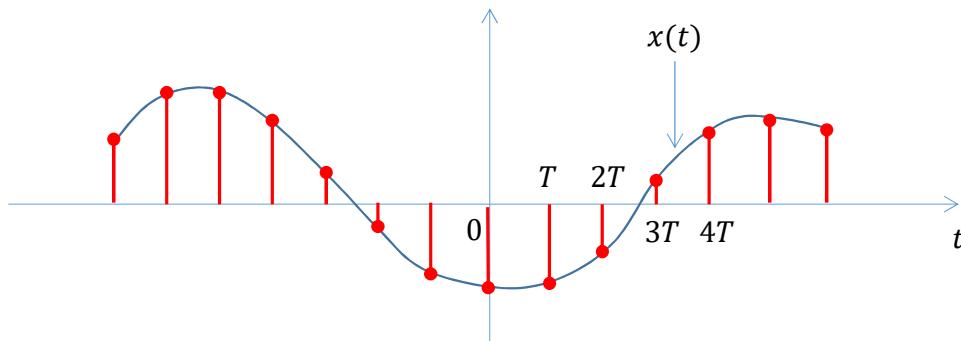
$$h(t) = 2 \frac{\omega_c}{\pi} \text{sinc}(\omega_c t)$$



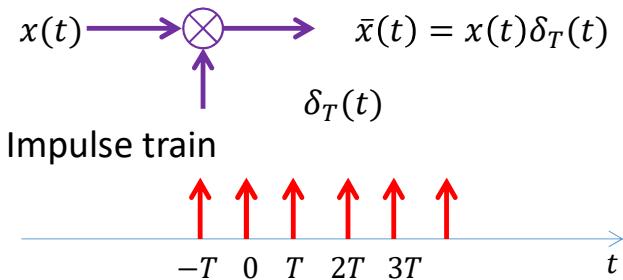
$\omega_c$ , denoted as the cut off freq of the lowpass filter has to be such that the original signal is fully recovered.  $B_0 \leq \omega_c$ ,  $\omega_c \leq 2\omega_0 - B_0$

The Amplitude Modulated (AM radio) carrier frequency  $\omega_0$  is in the frequency range 535-1605 kHz. Each carrier frequency is assigned at 10 kHz intervals.

## Sampling

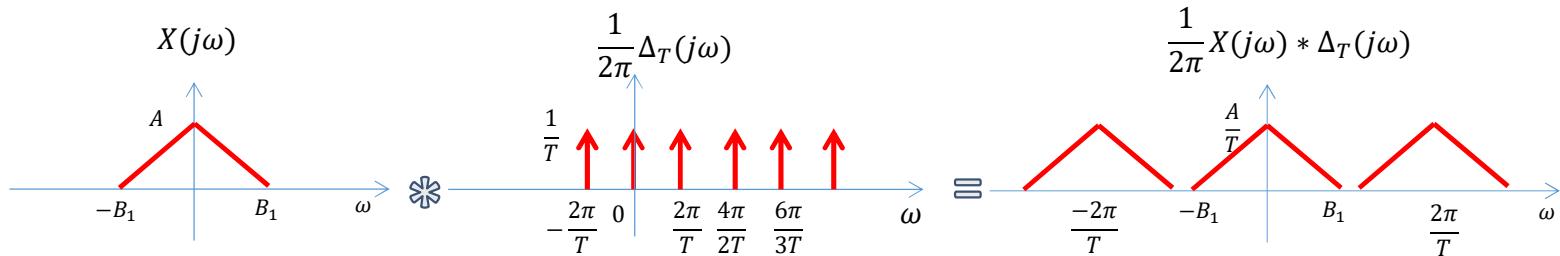


Analog input  $x(t)$  → Sampling at every  $T$  unit →  $x[n]$  Discrete output

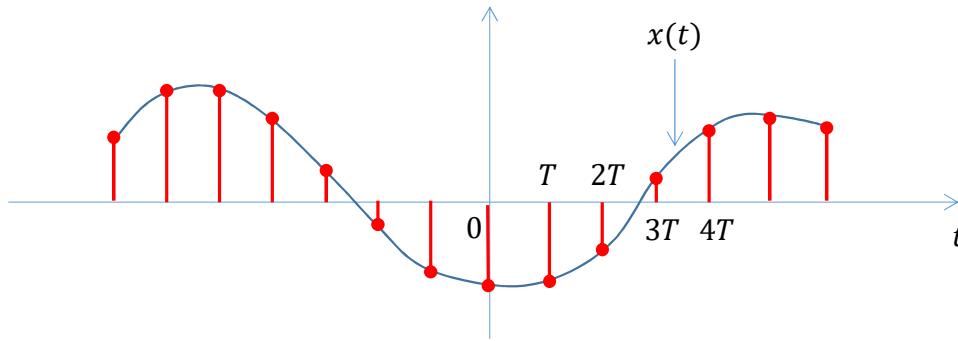


$$x(t)\delta_T(t) \xrightarrow{FT} \frac{1}{2\pi} X(j\omega) * \Delta_T(j\omega)$$

$$\begin{aligned}\bar{x}(t) &= x(t) \times \delta_T(t) \\ &= x(t) \sum \delta(t - nT) \\ &= \sum x(t)\delta(t - nT) \\ &= \sum x(nT)\delta(t - nT)\end{aligned}$$



## Sampling



Analog input  $x(t)$  → Sampling at every  $T$  unit →  $x[n]$  Discrete output

$$x(t) \xrightarrow{\otimes} \bar{x}(t) = x(t)\delta_T(t)$$

Impulse train

$$x(t)\delta_T(t) \xrightarrow{FT} \frac{1}{2\pi} X(j\omega) * \Delta_T(j\omega)$$

Sampling Theorem: Only if the sampling is fast enough ( $\frac{2\pi}{T} \geq 2B_1$ ), the original signal  $x(t)$  can be recovered from the sampled  $\bar{x}(t)$  through a lowpass filter.

